600 GHz Planar-Schottky - Diode Subharmonic Waveguide Mixers

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ABSTRACT

We report on the first planar two-diode subharmonic mixer operating at 600 GHz. The measured mixer noise temperature of 4200 K DSB is only a factor of 2 worse than the best single diode (whisker-contacted or planar) fundamental mixers and a factor 1.5 worse than the best ever single diode (whisker-contacted) harmonic mixer at similar frequencies. The two-diode subharmonic mixer offers the advantage of using a local oscillator at half the signal frequency and has a relatively wide achievable 1 F bandwidth. The measured conversion loss is ≈12 dB and a noise temperature between 4200 -S300 K can be maintained with a fixtuned circuit over a 1-18 GHz IF bandwidth. The mixer mount consists of standard height waveguides with an external signal feed horn and LO coupling via waveguide through a standard microstrip probe/filter. A single quartz substrate contains the LO and IF filters/coupling lines as well as the antiparallel-pair Schottky diodes. The diodes are integrated into the quartz substrates using a now] fabrication process which removes all of the unwanted GaAs from the regions containing the microstrip lines. The anodes are made via a trilayer direct c-beam write process that results in extremely low parasitic devices with high cutoft' frequencies. The mixers are being developed for use on Earth Observing System Microwave I imb Sounder, an ozone chemistry experiment which is part of NASA's Mission to Planet Earth.

I. INTRODUCTION

Planar Schottky dc.vices have been under development for a number of years sow, with the hope that they will replace the whisk-coat acted honeycomb diodes currently being used for most high frequency space-tm ne applications. Recent results have, indicated that the appropriate planar structure can perform as well as whisker contacted diodes at least up to 350 GHz [i-4]. Integrated circuit technology has also enabled the successful implementation of mixers and multipliers that require more than one diode.

Since all devices are made at the same time and in close proximity, the risk of large variations from junction to junction is greatly reduced. Currently, we are in tile process 01 developing flight-qualified 240 GHz and 640 GHz subharmonic mixers based on the anti-parallel-pair diode configuration [5] for use in the Earth Observing System Microwave 1 imb Sounder instrument, an ozone chemistry experiment which is part of NASA's Mission to Planet Liarth

In this paper, we report on the first subharmonic mixer working at 600" GHz which utilizes two planar diodes in an anti-parallel-pair configuration. The main advantages of this particular mixer configuration for Earth remote sensing applications, lie in the simple local oscillator diplexing (the 1.0 is em-half of the signal frequency and can be injected efficiently through waveguide without narrow band fillers to suppress 1.0 noise in the signal band), and the broad achievable 1 i bandwidth, which allows several spectral lines to be measured with a single fixed-frequency 1.0 (the He output impedance is between 80 and I 00 ohms making possible broadband matching to the IF amplifier). Moreover, this particular mixer configuration has a measur cd 1,0 noise suppression of better than 30 dB due to the anti-parallel-pair configuration[3]. At 200 GHz, using a scaled version of the waveguide mount we report upon here, we have measured an ambient input noise temperature [4] which is about 30 percent lower (baa that of the best whisker-contacted diode subharmonic mixers [5] and only a factor 01::1.5 worse than the, best ever fundamental mixer al this frequency [6]. At 614 GHz, currently, we are measuring a noise temperature of 4200 K with a conversion loss of about 12. dB. This is only a factor of 2 worse than the best reported single diode (whisker-contacted or planar) fundamental mixers [7,8,9] and a factor 1.5 worse than the best ever single diode (whisker-contacted) harmonic mixer [1 0] at similar frequencies. Section II describes the dc.vice technology that has been developed for this application and Section 111 discusses the measured mixer results in more detail.

11. FABRICATIONTECHNOLOGY

We have developed two major modifications to current submicron device fabrication technology that enabled us to produce very high frequency planar-Schottky-diode waveguide mixers. I first, we have, monolithically mated the GaAs diodes withthe lower-(liclcc[ric-co)~s! al]t lower-loss quartz-based llJiclow:lvc/lllillit llctcr-wave circuitry used to couple power in and out of the mixer block. This has been accomplished using a process we refer to as QUID (Quartz Up-side-dowa Integrated Device) [4]. In [his process, the completely integrated G a A s device and surrounding circuitry arc mounted up-side-down on a quartz carrier with the help of a thermally cured epoxy. The GaAs substrate is then Completely etched away everywhere except for a small region around the active devices. A final backside dry mesa etch ensures that all of the unnecessary GaAs is etched away (Figs. 1.2). After the backside process, individual circuits are diced and glued into their supplied microstrip housing in the waveguide block (Fig. 3). Two wire binds to the circuit metallization provide a 1)(: return on one side of the diodes and an IF output line to a standard K-connector bead on the other. The 600" GHz circuits are fabricated on 2cm diameter, 50 micronthick crystal quartz and after dicing, each circuit is about 80 microns wide by 250[) microns long.

The second modification we have introduced to the traditional planar-Schottky-diode fabrication process is the technique of using c-beam technology to directly write the anode, and the connecting finger in the same step. A trilayer I'MM A process was developed for this application which has yielded Schottky diodes with Iower-tll:ltl-average parasitic capacitance and series resistance [4], technology is similar to the '1'-gale technology that has been used for high frequency HEMT structures. An S1 3M picture of a typical T-anode diode, before. the addition of a surface channel for anode-to-cathode isolation, is shown in 1 figure 1. In addition to the process that has been devel oped and already described [4,11] for producing 200" G} lz diodes, for the 600" GHz (ic.vices the horizontal part of the "1" was raised anadditional 1 00-150 nm in order to further decrease the parasitic capacitance.

The 600" GHz waveguide mixer mount uses a traditional crmscd-guide split-block configuration with the local oscillator waveguide perpendicular to the signal guide and electrically coupled via a shielded quartz microstrip. An external feed horn is used for signal coupling while the 10 enters via waveguide. A single quartz substrate contains the LO and II filters and coupling lines as well as the antiparallel-pair Schottky diodes. Hammerhead filter elements are used for frequency separation and matching. A more detailed description of the block that was used al 215 GHz has been reported elsewhere [3]. The waveguide block used for the 600" GHz work is a scaled version of the 2.15 GHz design with some minor modifications to the input

signal portand I. O and IF filter structures. A descriptive schematic of the block is shown in Figure 3.

111. MEASUREMENTS AND DISCUSSION

The first batch of successful devices had an anode area of $0.8 \mu m^2$ (().4 X 2.0 \(\mu\mathrm{m}\)). The particular diode pair that resulted in the mist temperature reported here bas a series resistance of 5.6/5.8 Ohms, an ideality factor of 1.26/1.28 and a reverse saturation current of 4.8 x 1() /5.2 x 10. Amps pet anode respectively. 'J' he total measured capacitance of the diodes (including the microstrip circuit on the quartz substrate) is about I 3 ft. 1 from previous measurements, the microstrip contributes about 4 ft of capacitance and the anode capacitance is calculated to be about 1 to 1.5 ft. This leaves a total parasitic capacitance of 6-7 ft. We believe that this capacitance can be further reduced by small variations to the anode finger overlap and T- gate structure.

1 figure 4. shows the measured performance of the mixer over an I I; frequency band of 1 to 18 GHz. The tuners were optimized and fixed at 8 GHz 11'. The real part of the Houtput impedance, as measured 011 an 8510 network analyzer, is approximately 90 Ohms over the. whole band. The best performance was obtained at an II of 3 GHz with the measured mixer noise temperature of 4180 K DSB and a conversion loss of 11.9 dB. The required LO power for this particular measurement was excessively high (>5[) mW), however we have determined that most of the power is being lost in a poorly machined 1. O waveguide which cannot be corrected in the current mixer block. We have recently (in a newly fabricated mixer block) been able to turn on both devices with only 10mW of power at the mixer input 1. O port. We have also determined that there is additional 10 power 10ss in the microstrip coupling circuit between the waveguide probe and the diodes which we attribute to the thermal epoxy that binds the metallic film of the microstrip line to the quartz substrate. This 10ss can be reduced by reducing the thickness of the glue layer (currently >1 0µm). We expect future device runs to have a glue layer which is <2μm using a new curing technique we have been experimenting with. I for comparison put poses "1'able Ishows previously published results with roomtemperature Scholtky-diode mixers around 600" GHz. The present performance is about a factor of two worse than the results obtained from single-diode whisker-contacted or planar-Schottky-diode mixers. We expect our noise numbers to improve as both devices and mount are optimized. The harmonic single-diode mixer results reported in [10] are indeed very impressive and are about a factor of 1.5 better than tbc results from our subharmonic two-diode mixer. Compared to single-diode har monic mixers, the main advantage of the two-diode arrangement is a significantly lower output impedance and potentially wide.r IF bandwidth. Moreover, it should be noted that the harmonic mixer uses a whisker contacted diode which generally has lower parasitics.

IV. CONCLUSION

Stale-of-the-ar(mixer performance has been measured with an anti-parallel-pair-diode subharmonic mixer in a waveguide circuit at 600" GHz. The measured noise temperature and conversion 10ss are only a factor of 2 worse than the best ever fundamental mixers at similar frequentics. '1'0 the best of our knowledge, these are the first results from a subharmonic mixer with two diodes at 600 GHz. We are currently improving the quality of the devices and reducing the 10sscs in the waveguide mount and believe that even helter results will be forthcoming in the near future.

V. ACKNOWLEDGMENT

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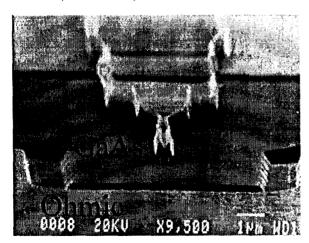


Fig. 1. SEM close-up of T-anode Schottky diode before the channel etch which isolates the anode and cathode.

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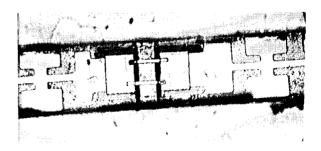
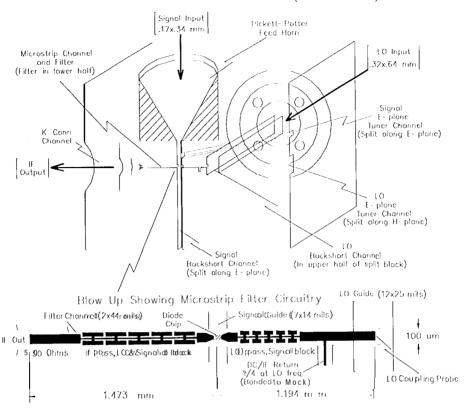
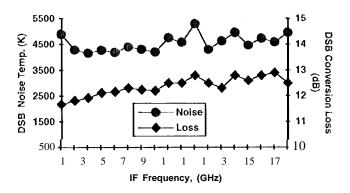


Fig. 2. Photo of the antiparallel-pair diodes and surrounding filter circuitry as seen through the quartz substrate.

640 GHz SHPM Mixer Block: Lower Half (19x19x9.5mm)





1 ig. 4. Measured performance of the subharmonic mixer at 614 GHz over 1-18 GHz IF. The average IF output impedance is 90 Ohms. The RF tuning was optimized at 8 GHz IF.

Freq GHz	Junction '1'ype	Mixer Type	I I .oss [(1)\$11) _	Tm DSB	Ret'.
665	Whisker	Harmonic Waveguide	~12 dB	2650 K	[10]
649	whisker	Fundamental Waveguide	8.5 dB	1900 K	[7]
760	planar I	Fundamental log-periodic	14.9 dB	89()() K	[8]
614	planar 1	Subharmonic Waveguide	12 dB	4200" K	This work
585	planar	Fundamental Waveguide	8.66 dB	2030" K	[9]

'1'able 1, Comparison of various devices and mixer circuits operating ≈ 600 G]17.